



6. Draw the block scheme of the following continuous-time system where with  $\mathbf{x}_c$  denotes the vector  $\mathbf{x}_c = [x_1 \ x_2 \ x_3 \ x_4]^T$ .

$$\begin{cases} \dot{\mathbf{x}}_c(t) = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -\alpha_0 & -\alpha_1 & -\alpha_2 & -\alpha_3 \end{bmatrix} \mathbf{x}_c(t) + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 1 \end{bmatrix} u(t) \\ y(t) = [ \beta_0 \ \beta_1 \ \beta_2 \ \beta_3 ] \mathbf{x}_c(t) \end{cases}$$

$$\boxed{\frac{1}{s}}$$

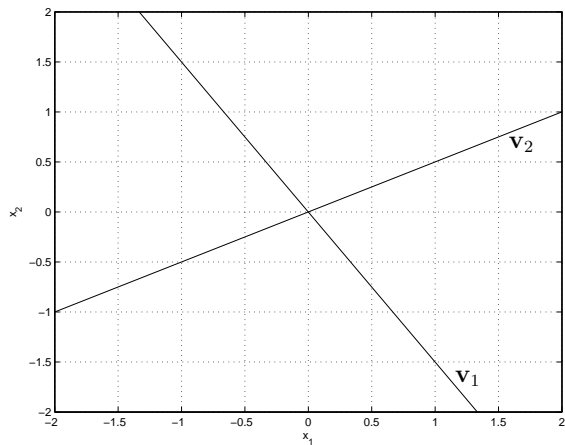
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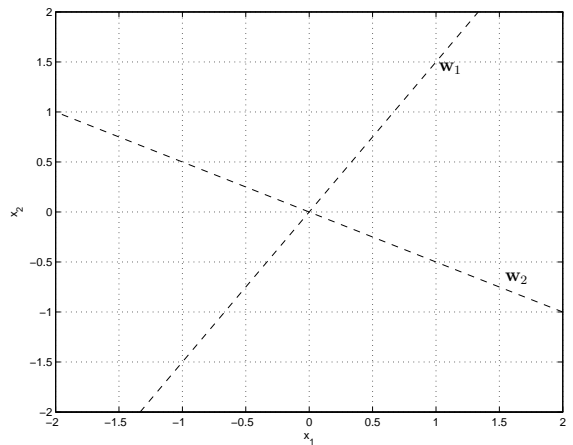
7. Draw qualitatively the trajectories of a second order dynamic system  $\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t)$  characterized by the eigenvalues  $\lambda_i$  and the eigenvectors  $\mathbf{v}_i$  shown in the two following boxes.

1) Eigenvalues:  $\lambda_1 = -1$  and  $\lambda_2 = -3$ . The corresponding real eigenvectors  $\mathbf{v}_1$  and  $\mathbf{v}_2$  are shown in figure.



- Node?     Focus?     Saddle?  
 Stable?     Unstable?

2) Eigenvalues:  $\lambda_{1,2} = 2 \pm 3j$ . Eigenvectors:  $\mathbf{v}_1 = \mathbf{w}_1 + \mathbf{w}_2j$  and  $\mathbf{v}_2 = \mathbf{v}_1^*$ . The real vectors  $\mathbf{w}_1$  and  $\mathbf{w}_2$  are shown in figure.



- Node?     Focus?     Saddle?  
 Stable?     Unstable?

8. Given the transfer function  $G(z)$ , write the structure of corresponding dynamic system in the reachability canonical form denoting with  $u(k)$  the input and with  $y(k)$  the output:

$$G(z) = \frac{z^4 + 6z^3 + 4z^2 + 7z + 4}{z^4 + 5z^3 + 2z^2 + 3z + 1} \quad \left\{ \begin{array}{l} \mathbf{x}(k+1) = \begin{bmatrix} \phantom{0} \\ \phantom{0} \\ \phantom{0} \\ \phantom{0} \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} \phantom{0} \\ \phantom{0} \\ \phantom{0} \\ \phantom{0} \end{bmatrix} u(k) \\ y(k) = [ \phantom{0} \ \phantom{0} \ \phantom{0} \ \phantom{0} ] \mathbf{x}(k) + [ \phantom{0} ] u(k) \end{array} \right.$$

9. Consider the point-to-point control problem for a continuous-time linear system. Write the necessary condition that must be verified such that it exists a control law  $\mathbf{u}$  that moves the system from the initial state  $\mathbf{x}(0)$  to the final state  $\mathbf{x}(t)$  in the time interval  $[0, t]$ :

...

10. Compute the  $k$ -th power of the following Jordan miniblock:

$$\begin{bmatrix} \lambda & 1 & 0 & 0 \\ 0 & \lambda & 1 & 0 \\ 0 & 0 & \lambda & 1 \\ 0 & 0 & 0 & \lambda \end{bmatrix}^k = \begin{bmatrix} & & & \\ & & & \\ & & & \\ & & & \end{bmatrix}$$

11. Given a SISO linear system of the fourth order, completely observable, characterized by matrices  $\mathbf{A}$ ,  $\mathbf{b}$  and  $\mathbf{c}$ . Let  $p(\lambda) = \lambda^4 + \alpha_3\lambda^3 + \alpha_2\lambda^2 + \alpha_1\lambda + \alpha_0$  be the characteristic polynomial of matrix  $\mathbf{A}$ . Write the structure of matrix  $\mathbf{P}$  which, together with the space transformation  $\mathbf{x} = \mathbf{P}\mathbf{x}_o$ , brings the system in the observability canonical form.

$$\mathbf{P} =$$

12. Given the continuous-time linear system  $\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t)$ , write the structure of:  
 a) an **open loop** state estimator and the time evolution of the corresponding estimation error  $\mathbf{e}(t) = \mathbf{x}(t) - \hat{\mathbf{x}}(t)$  starting from the initial condition  $\mathbf{e}(0)$ :

$$\dot{\hat{\mathbf{x}}}(t) = \quad \quad \quad \mathbf{e}(t) =$$

- b) a **in full order closed loop** state estimator and the time evolution of the corresponding estimation error  $\mathbf{e}(t) = \mathbf{x}(t) - \hat{\mathbf{x}}(t)$  starting from the initial condition  $\mathbf{e}(0)$ :

$$\dot{\hat{\mathbf{x}}}(t) = \quad \quad \quad \mathbf{e}(t) =$$

13. Relatively to the linear discrete system  $\mathbf{x}(k+1) = \mathbf{A}\mathbf{x}(k) + \mathbf{B}\mathbf{u}(k)$ ,  $\mathbf{y}(k) = \mathbf{C}\mathbf{x}(k)$ , write a necessary and sufficient condition for the complete “constructability” of the system:

14. Given the following continuous-time linear system  $\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t)$ ,  $\mathbf{y}(t) = \mathbf{C}\mathbf{x}(t)$ . Write the expression of the matrices  $\mathbf{F}$ ,  $\mathbf{G}$  and  $\mathbf{H}$  that characterize the corresponding sampled system  $\mathbf{x}(k+1) = \mathbf{F}\mathbf{x}(k) + \mathbf{G}\mathbf{u}(k)$ ,  $\mathbf{y}(k) = \mathbf{H}\mathbf{x}(k)$  with period  $T$ :

$$\mathbf{F} = \quad \quad \quad \mathbf{G} = \quad \quad \quad \mathbf{H} =$$

The corresponding sampled system is reachable if and only if for each couple  $\lambda_i, \lambda_j$  of eigenvalues of  $\mathbf{A}$  having the same real part it is:

...

15. Write the Ackermann formula for computing the gain vector  $\mathbf{l}$  of an asymptotic state observer which freely places the eigenvalues of matrix  $\mathbf{A} + \mathbf{l}\mathbf{c}$ :

$$\mathbf{l} =$$

This formula can be used

- only for systems with one input;
- only for systems which are completely observable;
- for any type of time-invariant linear system;

16. Write the block matrices  $\overline{\mathbf{A}}$ ,  $\overline{\mathbf{B}}$  and  $\overline{\mathbf{C}}$  of a system in the reachability standard form:

$$\overline{\mathbf{A}} = \begin{bmatrix} & \\ & \end{bmatrix}, \quad \overline{\mathbf{B}} = \begin{bmatrix} \\ \end{bmatrix}, \quad \overline{\mathbf{C}} = \begin{bmatrix} & \end{bmatrix}$$

17. Given a discrete-time nonlinear system  $\mathbf{x}(k+1) = \mathbf{f}(\mathbf{x}(k))$ . The equilibrium points of this system can be obtained :

- computing the points  $\mathbf{x}_e(k)$  tali that  $\mathbf{x}_e(k) = \mathbf{f}(\mathbf{x}_e(k))$ .
- solving the equation  $0 = \mathbf{f}(\mathbf{x}_e(k))$  with respect to the unknown  $\mathbf{x}_e(k)$ .
- computing the points of the state space where  $\mathbf{x}_e(k+1) = \mathbf{x}_e(k)$ .

18. Write the structure of the dual system  $\mathcal{S}_D$  corresponding to a given system  $\mathcal{S} = (\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D})$ :

$$\mathcal{S}_D = ( \quad , \quad , \quad , \quad )$$

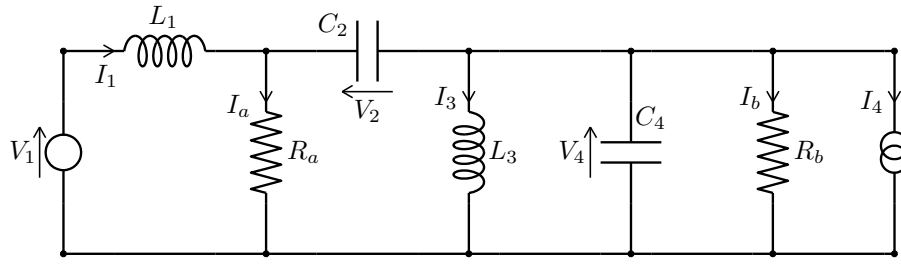
19. Consider a continuous-time nonlinear system  $\dot{\mathbf{x}}(t) = \mathbf{f}(\mathbf{x}(t), \mathbf{u}(t))$ ,  $\mathbf{y}(t) = \mathbf{g}(\mathbf{x}(t), \mathbf{u}(t))$  and let  $\mathbf{x}_0$  be an equilibrium point of the system for constant input  $\mathbf{u}_0$ . Write the definition of the matrices  $\mathbf{A}$ ,  $\mathbf{B}$ ,  $\mathbf{C}$ ,  $\mathbf{D}$  of the linearized system:

$$\mathbf{A} = \quad \quad \quad \mathbf{B} = \quad \quad \quad \mathbf{C} = \quad \quad \quad \mathbf{D} =$$

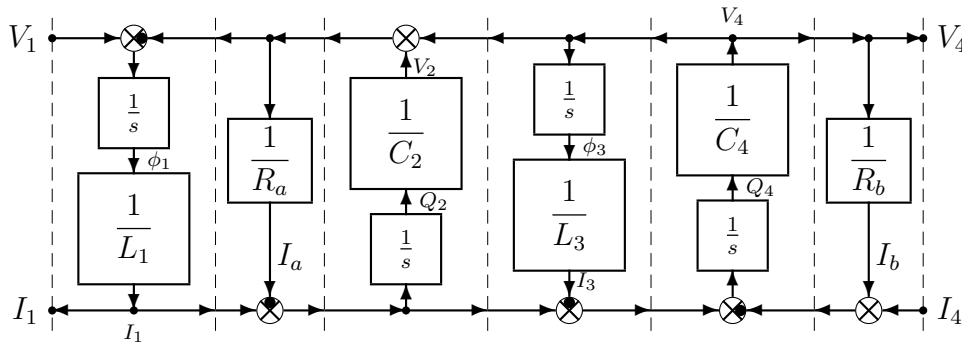
20. Write, within the following table, the symbols and the names of the energy variables and the power variables that characterize the Energetic Domain: *Mechanical Rotational*. Moreover, write the constitutive relation (both linear and non linear) and the differential equation which characterize the physical elements:

	Symbols / Names	Constitutive Rel.	Linear Case	Differential Eq.
$\mathcal{D}_1$				
$q_1$				
$v_1$				
$\mathcal{D}_2$				
$q_2$				
$v_2$				
$\mathcal{R}$				

21. Consider the following electric circuit composed by the inductances  $L_1$ ,  $L_3$ , the capacities  $C_2$ ,  $C_4$  and the resistances  $R_a$  and  $R_b$ . Two inputs act on the system: the voltage  $V_1$  and the current  $I_4$ . The outputs of the system are: the current  $I_1$  and the voltage  $V_4$ .



The POG model of the given electric circuit has the following structure:



Let  $\mathbf{x} = [I_1 \ V_2 \ I_3 \ V_4]^T$  be the state vector,  $\mathbf{u} = [V_1 \ I_4]^T$  the input vector and  $\mathbf{y} = [I_1 \ V_4]^T$  the output vector. Write the corresponding dynamic system  $\bar{\mathbf{L}}\dot{\mathbf{x}} = \bar{\mathbf{A}}\mathbf{x} + \bar{\mathbf{B}}\mathbf{u}$  and  $\mathbf{y} = \bar{\mathbf{C}}\mathbf{x} + \bar{\mathbf{D}}\mathbf{u}$  in the state space:

$$\underbrace{\begin{bmatrix} \phantom{0} \\ \phantom{0} \\ \phantom{0} \\ \phantom{0} \end{bmatrix}}_{\bar{\mathbf{L}}} \underbrace{\begin{bmatrix} \dot{I}_1 \\ \dot{V}_2 \\ \dot{I}_3 \\ \dot{V}_4 \end{bmatrix}}_{\dot{\mathbf{x}}} = \underbrace{\begin{bmatrix} \phantom{0} \\ \phantom{0} \\ \phantom{0} \\ \phantom{0} \end{bmatrix}}_{\bar{\mathbf{A}}} \underbrace{\begin{bmatrix} I_1 \\ V_2 \\ I_3 \\ V_4 \end{bmatrix}}_{\mathbf{x}} + \underbrace{\begin{bmatrix} \phantom{0} \\ \phantom{0} \\ \phantom{0} \\ \phantom{0} \end{bmatrix}}_{\bar{\mathbf{B}}} \underbrace{\begin{bmatrix} V_1 \\ I_4 \end{bmatrix}}_{\mathbf{u}}$$

$$\underbrace{\begin{bmatrix} I_1 \\ V_4 \end{bmatrix}}_{\mathbf{y}} = \underbrace{\begin{bmatrix} \phantom{0} \\ \phantom{0} \end{bmatrix}}_{\bar{\mathbf{C}}} \mathbf{x} + \underbrace{\begin{bmatrix} \phantom{0} \\ \phantom{0} \end{bmatrix}}_{\bar{\mathbf{D}}} \underbrace{\begin{bmatrix} V_1 \\ I_4 \end{bmatrix}}_{\mathbf{u}}$$

22. Write the La Salle - Krasowskii stability criterion for discrete-time nonlinear systems.

Consider the discrete-time nonlinear system  $\mathbf{x}(k+1) = \mathbf{f}(\mathbf{x}(k), \mathbf{u}_0)$  and let  $\mathbf{x}_0$  be an equilibrium point corresponding to the constant input  $\mathbf{u}_0$ .

If ...

23. Given the following continuous-time nonlinear system  $\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, u)$ :

$$\begin{cases} \dot{x}_1 &= x_2^2 - x_1^3 - x_1 \\ \dot{x}_2 &= \alpha x_2 - x_1^3 x_2 + u \end{cases}$$

Set  $u = 0$ , it is easy to verify that the origin  $\mathbf{x}_1 = (0, 0)$  is an equilibrium point for the system.

a) Compute the matrix  $\mathbf{A}(\mathbf{x}) = \frac{\partial \mathbf{f}(\mathbf{x}, u)}{\partial \mathbf{x}}$  and the vector  $\mathbf{b} = \frac{\partial \mathbf{f}(\mathbf{x}, u)}{\partial u}$  of the nonlinear system:

$$\mathbf{A}(\mathbf{x}) = \begin{bmatrix} & \\ & \end{bmatrix}, \quad \mathbf{b} = \begin{bmatrix} \\ \end{bmatrix}$$

b) Compute the matrix  $\mathbf{A}_1$  and the vector  $\mathbf{b}_1$  of the linearized system at the point  $\mathbf{x}_1 = (0, 0)$ ,  $u = 0$ :

$$\mathbf{A}_1 = \begin{bmatrix} & \\ & \end{bmatrix}, \quad \mathbf{b}_1 = \begin{bmatrix} \\ \end{bmatrix},$$

c) Study, for varying parameter  $\alpha$ , the stability of the nonlinear system in the neighborhood of point  $\mathbf{x}_1 = (0, 0)$ ,  $u = 0$  using the reduced Lyapunov criterion:

d) For  $\alpha = 0$  and  $u = 0$ , study the stability of the nonlinear system in the vicinity of the origin  $\mathbf{x}_1 = (0, 0)$  using the “direct” Lyapunov criterion and the function:  $V(\mathbf{x}) = x_1^4 + 2x_2^2$ . Eventually, use the La Salle - Krasowskii criterion.

e) Set  $\alpha = 0$  and  $u = k_1 x_1 + k_2 x_2$ , compute for which values of the parameters  $k_1$  and  $k_2$  the linearized system in the neighborhood of point  $\mathbf{x}_1 = (0, 0)$  is asymptotically stable.

